Lecture 28: Design of IIR Filters

Foundations of Digital Signal Processing

Outline

- Designing IIR Filters with Discrete Differentiation
- Designing IIR Filters with Impulse Invariance
- Designing IIR Filters with the Bilinear Transform
- Related Analog Filters

Designing IIR Filters

- No easy ways to design digital IIR filters
- So let us start from analog filters

Designing IIR Filters

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Option 1: Preserve the difference equation!

Question: What is a derivative in discrete-time?

In continuous-time

$$\frac{dx(t)}{dt} \to sX(s)$$

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$$\frac{dx(t)}{dt} \to sX(s)$$

$$\frac{dx(t)}{dt} = \lim_{\Delta T \to 0} \frac{x(t) - x(t - \Delta T)}{\Delta T}$$

$$\frac{dx(t)}{dt} \bigg|_{t=nT} = \frac{x(nT) - x(nT - T)}{T} = x[n] - x[n-1]$$

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$$\frac{dx(t)}{dt} \bigg|_{t=nT} \to \frac{1}{T} (1 - z^{-1}) X(z)$$

Question: What is a second-derivative in discrete-time?

In continuous-time

$$\frac{d^2 x(t)}{dt^2} \to s^2 X(s)$$

$$\frac{d^2 x(t)}{dt^2} = \frac{dx(t)}{dt} \left[\frac{dx(t)}{dt} \right]$$
$$\frac{dx(t)}{dt} \bigg|_{t=nT} = \frac{x(nT) - x(nT - T)}{T}$$

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$$\frac{d^2 x(t)}{dt^2} \bigg|_{t=nT} = \frac{[x(nT) - x(nT - T)]/T - [x(nT - T) - x(nT - 2T)]/T}{T}$$

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$$\frac{d^2 x(t)}{dt^2} = \frac{dx(t)}{dt} \left[\frac{dx(t)}{dt} \right]$$
$$\frac{d^2 x(t)}{dt^2} \bigg|_{t=nT} = \frac{x(nT) - 2x(nT - T) + x(nT - 2T)}{T^2} \to \frac{x[n] - 2x[n-1] + x[n-2]}{T^2}$$

Question: What is a second-derivative in discrete-time?

In continuous-time

$$\frac{d^2 x(t)}{dt^2} \to s^2 X(s)$$

$$\begin{aligned} \frac{d^2 x(t)}{dt^2} &= \frac{dx(t)}{dt} \left[\frac{dx(t)}{dt} \right] \\ \frac{d^2 x(t)}{dt^2} \bigg|_{t=nT} &= \frac{x(nT) - 2x(nT - T) + x(nT - 2T)}{T^2} \to \frac{x[n] - 2x[n-1] + x[n-2]}{T^2} \\ \frac{dx(t)}{dt} \bigg|_{t=nT} \to \frac{1}{T^2} (1 - 2z^{-1} + z^{-2}) X(z) = \frac{1}{T^2} (1 - z^{-1})^2 X(z) \end{aligned}$$

Question: What is a derivative in discrete-time?

Translate continuous-time to discrete-time

$$\frac{\frac{d^k x(t)}{dt^k} \to s^k X(s)}{\frac{d^k x(t)}{dt^k}} \xrightarrow{k} \frac{1}{T} (1 - z^{-1})^k X(z)$$

Question: What is a derivative in discrete-time?

Translate continuous-time to discrete-time

$$\frac{\frac{d^k x(t)}{dt^k}}{\frac{d^k x(t)}{dt^k}} \xrightarrow{} s^k X(s)$$

$$\frac{\frac{d^k x(t)}{dt^k}}{\frac{d^k x(t)}{dt^k}} \xrightarrow{} \frac{1}{T} (1 - z^{-1})^k X(z)$$

$$s \to \frac{1}{T}(1-z^{-1})$$

Example:
$$s \rightarrow \frac{1}{T}(1-z^{-1})$$

$$H(s) = \frac{1}{(s+0.1)^2 + 9}$$

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$$H(s) = \frac{1}{(s+0.1)^2 + 9}$$

$$H(z) = \frac{1}{\left(\frac{1}{T}(1-z^{-1})+0.1\right)^2 + 9} = \frac{T^2}{T^2 \left[\left(\frac{1}{T}(1-z^{-1})+0.1\right)^2 + 9\right]}$$

$$= \frac{T^2}{\left((1-z^{-1})+0.1T\right)^2 + 9T^2} = \frac{T^2}{\left((1+0.1T)-z^{-1}\right)^2 + 9T^2}$$

Example:
$$s \rightarrow \frac{1}{T}(1-z^{-1})$$

$$H(s) = \frac{1}{(s+0.1)^2 + 9}$$

$$H(z) = \frac{T^2}{((1+0.1T) - z^{-1})^2 + 9T^2}$$

$$((1+0.1T) - z^{-1})^2 + 9T^2 = 0$$

$$((1+0.1T) - z^{-1})^2 = -9T^2$$

$$(1+0.1T) - z^{-1} = \pm 3Tj$$

$$z^{-1} = (1+0.1T) \mp 3Tj$$

$$z = \frac{1}{1+(0.1 \mp 3j)T}$$

Example:
$$s \rightarrow \frac{1}{T}(1-z^{-1})$$

$$z = \frac{1}{1 + (0.1 + 3j)T}$$
 Poles

• Example:
$$s \to \frac{1}{T}(1 - z^{-1})$$

 $z = \frac{1}{1 + (0.1 + 3j)T}$



Question: What is a derivative in discrete-time?

Translate continuous-time to discrete-time

$$s \to \frac{1}{T}(1-z^{-1})$$

Pros:

Relatively simple

Cons:

- Very limiting
- Stable continuous-time poles can only be mapped to low frequencies



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Option 2: Preserve the impulse response!

Question: How else can I represent my transfer function?

Partial-fraction decomposition

$$H(s) = \sum_{k=1}^{K} \frac{c_k}{s - p_k}$$

Question: How else can I represent my transfer function?

$$H(s) = \sum_{k=1}^{K} \frac{c_k}{s - p_k}$$

$$h(t) = \sum_{k=1}^{K} c_k e^{p_k t}$$

Question: How else can I represent my transfer function?

$$H(s) = \sum_{k=1}^{K} \frac{c_k}{s - p_k}$$

$$h(t) = \sum_{k=1}^{K} c_k e^{p_k t}$$

$$h(nT) = h[n] = \sum_{k=1}^{K} c_k e^{p_k nT} = \sum_{k=1}^{K} c_k [e^{p_k T}]^n$$

Question: How else can I represent my transfer function?

$$H(s) = \sum_{\substack{k=1 \ K}}^{K} \frac{c_k}{s - p_k} = \sum_{\substack{k=1}}^{K} c_k e^{p_k t}$$
$$h(t) = \sum_{\substack{k=1}}^{K} c_k e^{p_k t}$$
$$h(nT) = h[n] = \sum_{\substack{k=1}}^{K} c_k e^{p_k nT} = \sum_{\substack{k=1}}^{K} c_k [e^{p_k T}]^n$$
$$H(z) = \sum_{\substack{k=1}}^{K} \frac{c_k}{1 - e^{p_k T} z^{-1}}$$

• Example:
$$H(z) = \sum_{k=1}^{K} \frac{c_k}{1 - e^{p_k T_z - 1}}$$

 Use impulse invariance to transform the following biquad filter into the discrete-time domain.

$$H(s) = \frac{1}{(s+0.1)^2 + 9}$$

• Example:
$$H(z) = \sum_{k=1}^{K} \frac{c_k}{1 - e^{p_k T_z - 1}}$$

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Poles:

$$(s + 0.1)^2 + 9 = 0$$

 $s = \pm 3j - 0.1$

• Example:
$$H(z) = \sum_{k=1}^{K} \frac{c_k}{1 - e^{p_k T_z - 1}}$$

 Use impulse invariance to transform the following biquad filter into the discrete-time domain.

$$H(s) = \frac{1}{(s+0.1)^2 + 9} = \frac{1/2}{s+3j+0.1} + \frac{1/2}{s-3j+0.1}$$

Poles:

$$(s + 0.1)^2 + 9 = 0$$

 $s = \pm 3j - 0.1$

• Example:
$$H(z) = \sum_{k=1}^{K} \frac{c_k}{1 - e^{p_k T_z - 1}}$$

 Use impulse invariance to transform the following biquad filter into the discrete-time domain.

$$H(s) = \frac{1}{(s+0.1)^2 + 9} = \frac{1/2}{s+3j+0.1} + \frac{1/2}{s-3j+0.1}$$

$$p_k = \pm 3j - 0.1$$

$$H(s) \to h(t) \to h(nT) \to H(z)$$

$$H(z) = \frac{1/2}{1 - e^{(-3j-0.1)T}z^{-1}} + \frac{1/2}{1 - e^{(3j-0.1)T}z^{-1}}$$

$$p_k$$

Example:
$$H(z) = \sum_{k=1}^{K} \frac{c_k}{1 - e^{p_k T} z^{-1}}$$

 $H(z) = \frac{1/2}{1 - e^{(-3j - 0.1)T} z^{-1}}$
 $+ \frac{1/2}{1 - e^{(3j - 0.1)T} z^{-1}}$



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Option 3: Preserve the definition of z!

Question: How are z and s related?

From continuous-time to discrete-time

$$s = j\Omega$$

 $e^{st} = e^{snT} = z^n$
 $z = e^{sT}$
Taylor Series
Expansion / Approximation

K

• Building an approximation ($e^x \approx 1 + x$)

$$z = \frac{e^{\frac{ST}{2}}}{e^{-\frac{ST}{2}}} \approx \frac{1 + sT/2}{1 - sT/2}$$

Question: How are z and s related?

From continuous-time to discrete-time

$$s = j\Omega$$

$$e^{st} = e^{snT} = z^n$$

$$z = e^{sT} \qquad s = \frac{1}{T}\ln(z)$$

Building an approximation

$$s \approx \frac{2}{T} \frac{z-1}{z+1}$$

Bilinear Expansion / Approximation

The Bilinear Transform

Continuous-time to discrete-time

$$s \to \frac{2}{T} \frac{1 - z^{-1}}{1 + z^{-1}}$$

Discrete-time to continuous-time

$$z \to \frac{1 + sT/2}{1 - sT/2}$$

• Example:
$$s \to \frac{2}{T} \frac{1-z^{-1}}{1+z^{-1}}, \quad z \to \frac{1+sT/2}{1-sT/2}$$

 Use the bilinear transform to transform the following biquad filter into the discrete-time domain.

$$H(s) = \frac{1}{(s+0.1)^2 + 9}$$

• Example:
$$s \to \frac{2}{T} \frac{1-z^{-1}}{1+z^{-1}}, \quad z \to \frac{1+sT/2}{1-sT/2}$$

 Use the bilinear transform to transform the following biquad filter into the discrete-time domain.

$$H(s) = \frac{1}{(s+0.1)^2 + 9}$$

$$H(z) = \frac{1}{\left(\frac{2}{T}\frac{1-z^{-1}}{1+z^{-1}} + 0.1\right)^2 + 9}$$

$$= \frac{(1+z^{-1})^2}{\left(\frac{2}{T}(1-z^{-1}) + 0.1\right)^2 + 9(1+z^{-1})^2}$$

Example:
$$s \to \frac{2}{T} \frac{1-z^{-1}}{1+z^{-1}}, \quad z \to \frac{1+sT/2}{1-sT/2}$$

$$H(z) = \frac{(1+z^{-1})^2}{\left(\frac{2}{T}(1-z^{-1})+0.1\right)^2 + 9(1+z^{-1})^2}$$



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Multi-pole Filters

- **Butterworth:** Maximally flat passband
- **Chebyshev:** Faster cutoff with passband ripple
- Elliptic: Fastest cutoff with passband and stopband ripple



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Butterworth Filter

Butterworth Filter of order N

$$|H(j\omega)| = \frac{1}{\sum_{k=1}^{N} (s - s_k)}$$
 $s_k = e^{\frac{j(2k+N-1)\pi}{2N}}$

Butterworth Filter

Butterworth Filter of order N

 N equally spaced poles on a circle on the left-hand-side of the splane



Foundations of Digital Signal Processing Lecture 28: Designing IIR Filters

Butterworth Filter

Properties of the Butterworth Filter

- It is maximally flat at $\omega = 0$
- It has a cutoff frequency $|H(\omega)| = \frac{1}{\sqrt{2}}$ at $\omega = \omega_c$
- For large *n*, it becomes an ideal filter

Chebyshev Filter

Chebyshev Filter of order N

$$|H(j\omega)| = \frac{1}{\sqrt{1 + \epsilon^2 C_n^2(\omega)}}$$

 $C_n^2(\omega)$ is an nth-order Chebyshev Polynomial

 ϵ^2 controls ripple

Chebyshev Filter

Chebyshev Filter of order N



Foundations of Digital Signal Processing Lecture 28: Designing IIR Filters

Chebyshev Filter

Properties of the Chebyshev Filter

- It has ripples in the passband and is smooth in the stopband.
- The ratio between the maximum and minimum ripples in the passband is

$$(1 + \epsilon^2) - 1/2$$

- If e is reduced (i.e., the ripple size is reduced), then the stopband attenuation is reduced.
- It has a sharper cut-off than a Butterworth filter, but at the expense of passband rippling

Elliptic Filter

Elliptic Filter of order N

$$|H(j\omega)| = \frac{1}{\sqrt{1 + \epsilon^2 R_n^2(\omega)}}$$

 $R_n^2(\omega)$ is an nth-order elliptic function

 ϵ^2 controls ripple

Elliptic Filter

Elliptic Filter of order N



Foundations of Digital Signal Processing Lecture 28: Designing IIR Filters

Elliptic Filter

Properties of the Elliptic Filter

- It has ripples in the passband and the stopband
- The ratio between the maximum and minimum ripples is larger than the Chebyshev filter, but it has an even quicker transition from passband to stopband
- It has poles and zeros, but they are much more difficult to compute compared with the Butterworth and Chebyshev filters